

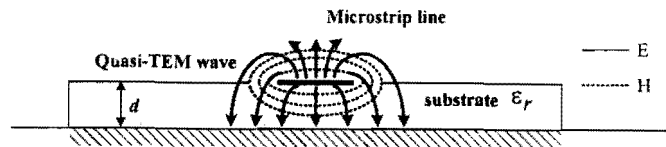
※ 考生請注意：本試題可使用計算機，並限「考選部核定之國家考試電子計算器」機型

* $\epsilon_0 = 10^{-9}/36\pi$; $\mu_0 = 4\pi \times 10^{-7}$; $\eta_0 = 120\pi$

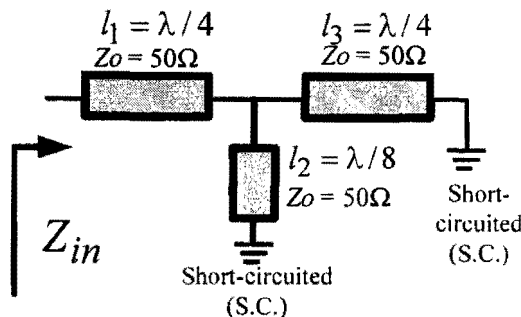
1. (a) Briefly explain why the *microstrip line can not support pure TEM-wave*. (5%)
- (b) For **50-Ω** microstrip line is on a substrate with $\epsilon_r = 4$ and $d = 8 \mu\text{m}$ at $f = 60 \text{ GHz}$, use the **quasi-TEM-wave approximation**, determine the guided wavelength λ_{eff} and wave velocity v_p in the microstrip line. (10%)

$$\epsilon_{\text{eff}} = \frac{\epsilon_r + 1}{2} + \frac{\epsilon_r - 1}{2} \frac{1}{\sqrt{1 + 12(d/W)}}$$

* Use the **pure TEM-wave approximation** to the required line width W



2. As shown in the following, determine the input impedance Z_{in} of the transmission network (10%)



$$Z_i = Z_0 \frac{Z_L + jZ_0 \tan(\beta l)}{Z_0 + jZ_L \tan(\beta l)}$$

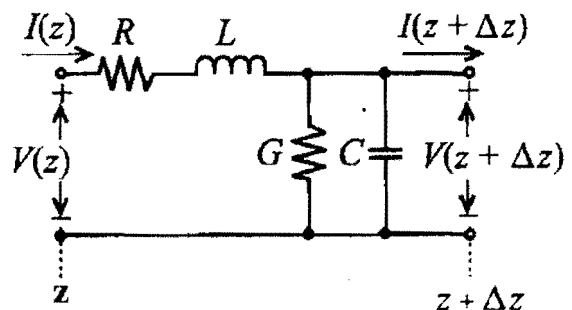
3. The **propagation constant** of a lossy transmission line is

$$\gamma (= \alpha + j\beta) = \sqrt{(R + j\omega L)(G + j\omega C)}$$

It is called the **distortionless line** when

$$R/L = G/C$$

- (a) Determine α , β , and phase velocity v_p of a **distortionless line**. (10%)
- (b) From β , explain the advantage of a **distortionless line**. (5%)



(背面仍有題目,請繼續作答)

系所組別： 電腦與通信工程研究所丙組

考試科目： 電磁學及電磁波

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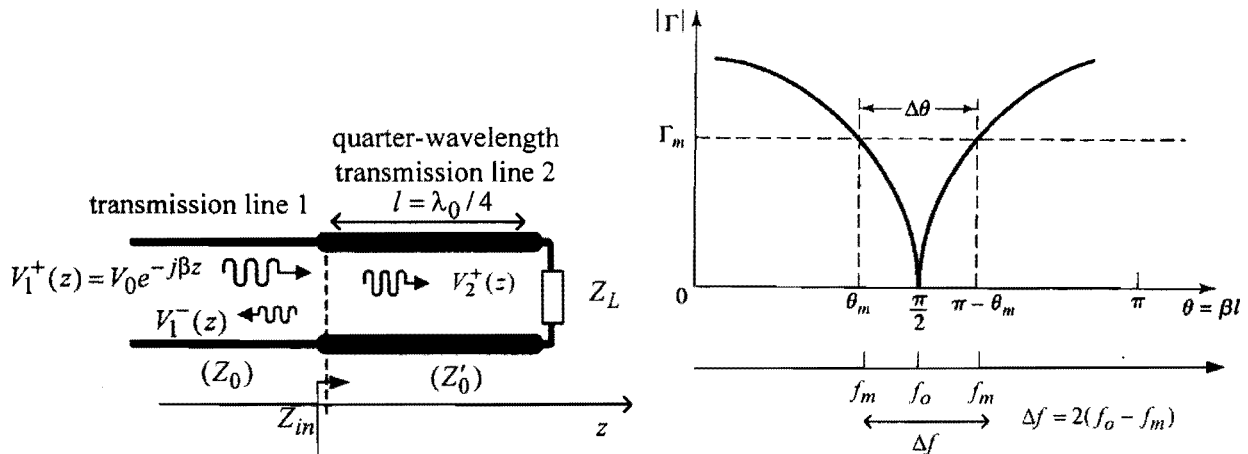
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4. For a quarter-wavelength line transformer with a real load impedance (Z_L), the following curve shows an approximate response of the reflection coefficient Γ near the central frequency (f_0). If $\Delta\theta \ll \pi/2$ and $\Delta f \ll f_0$. Γ_m can be derived as

$$\Gamma_m^{-2} = 1 + \left[\frac{2\sqrt{Z_0 Z_L} \sec\theta}{(Z_L - Z_0)} \right]^2$$

(a) Prove that
$$\frac{\Delta f}{f_0} = 2 - \frac{4}{\pi} \cos^{-1} \left[\frac{\Gamma_m}{\sqrt{1 - \Gamma_m^2}} \frac{2\sqrt{Z_0 Z_L}}{|Z_L - Z_0|} \right] \quad (5\%)$$

- (b) For $\Gamma_m = 0.5$ and $Z_L = 4Z_0$, determine Δf and **reflected & transmitted voltage** (to the load), $V_1^-(z)$ & $V_2^+(z)$, at $f = f_m$. (10%)



5. An **automotive tunnel** (汽車隧道) with a rectangular cross section (*width* $a = 15\text{m}$ & *height* $b = 6\text{m}$) has **aluminum metallic walls** ($\sigma_c = 4 \times 10^8$). If this tunnel is treated as a **waveguide** :

- (a) Determine the **lowest frequency** of the radio wave that will propagate through this tunnel and write down the mode (TE_{mn} or TM_{mn}) of this wave. (5%)
 (b) Let the **tunnel length** is **100 m** and a **12-MHz radio wave** propagating into this tunnel. Find the **total attenuation** (dB) of this radio wave through this tunnel. (5%)
 (c) Determine the **VSWR** of the radio wave at the end of the tunnel-waveguide (like having a free-space load). (5%)

Note: Waveguide wavelength and TE_{10} -mode impedance

$$\lambda_g = \lambda / \sqrt{1 - (f_c/f)^2} \quad \& \quad Z_{TE} = \eta_0 / \sqrt{1 - (f_c/f)^2}$$

Waveguide TE_{10} mode attenuation constant :

$$\alpha_{cTE10} = \frac{\lambda}{b\lambda_g} \sqrt{\frac{\pi}{\lambda\eta_0\sigma_c}} \left[1 + \left(\lambda_g / \lambda_c \right)^2 \left(1 + 2 \frac{b}{a} \right) \right]$$

系所組別： 電腦與通信工程研究所丙組

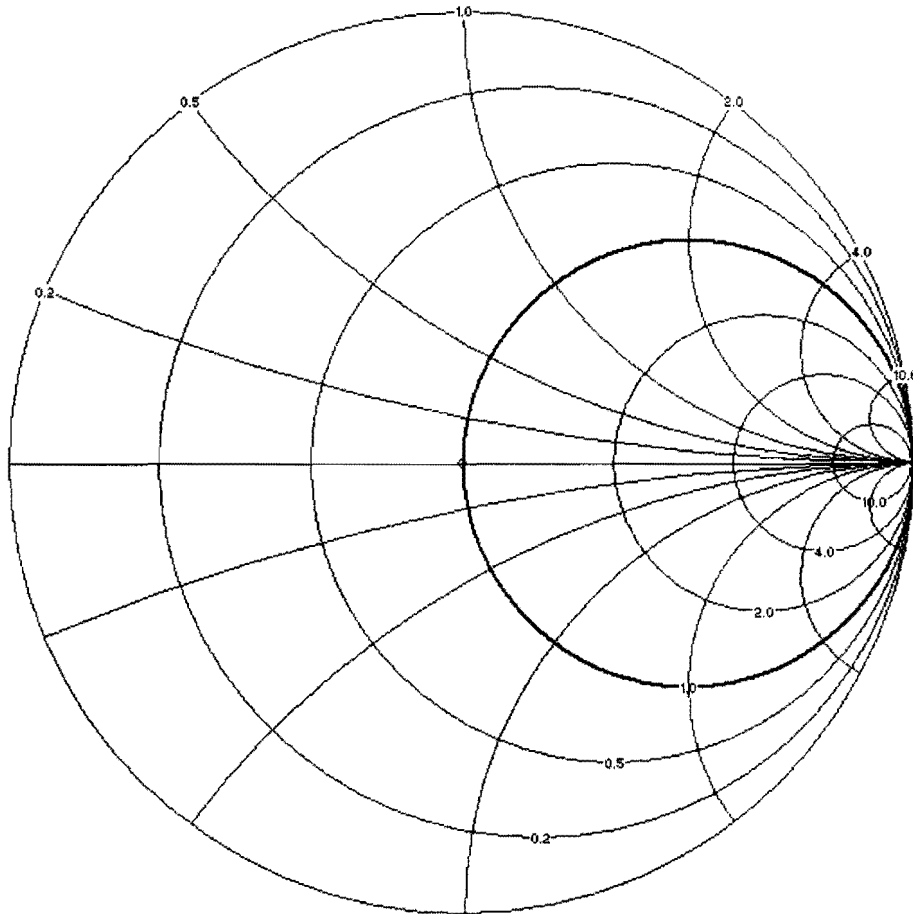
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6. Plot the **normalized impedance point** (to 50 Ω) in the Smith Chart and the corresponding **SWR circles** (also calculate the **SWR value**). (10%)

(a) $Z_L = 50 \Omega$; (b) $Z_L = 75 - j25 \Omega$; (c) $Z_L = 0 \Omega$; (d) $\Gamma_L = 0.5 \angle 45^\circ$



7. Let the transmitter output power is $P_t = 1 \text{ W}$ and the dipole-antenna input and transmitter output impedance are shown in the figure ($f = 10 \text{ GHz}$).

(a) Calculate the relected power (to the transmitter) and dipole-antenna radiation power P_{rad} (dBm) (if the antenna **radiation efficiency** $\eta_{rad} = 0.8$). (10%)

(b) From the antenna directive gain pattern (in dB) shown in the figure, calculate the **radiation power density** (p_{av}) and **E-field strength** in the direction of $\theta = 60^\circ$ and distance $r = 1 \text{ m}$. (10%)

