

# 國立中山大學 107 學年度碩士暨碩士專班招生考試試題

科目名稱：通訊理論【通訊所碩士班甲組、乙組選考】

題號：437002

※本科目依簡章規定「可以」使用計算機（廠牌、功能不拘）（問答申論題） 共 3 頁第 1 頁

1. (10%) A double-sideband amplitude modulation (AM) transmitted signal can be expressed as

$$u(t) = 2[1 + \pi \sin 2\pi t] \cos(2\pi f_c t + \phi_c),$$

where  $f_c$  is the carrier frequency and  $\phi_c$  is the phase. Can a simple envelope detector perfectly detect the message of  $\pi \sin 2\pi t$ ? Explain your answer.

2. (10%) A binary PSK demodulation and detection receiver with a carrier phase error  $\phi_e$  is considered, and the error probability can be expressed as

$$P(\phi_e) = Q\left(\sqrt{\frac{2E_b}{N_0} \cos^2 \phi_e}\right),$$

where  $Q(x) = \frac{1}{\sqrt{2\pi}} \int_x^\infty \exp\left(-\frac{u^2}{2}\right) du$ . Assume that the probability density function (PDF) of  $\phi_e$  is

$$p(\phi_e) = \frac{1}{\sqrt{2\pi}\sigma_{\phi_e}} e^{-\frac{\phi_e^2}{2\sigma_{\phi_e}^2}}. \text{ Decide the expression for the average error probability in an integral form.}$$

3. (25%) If two equiprobable messages of  $s_1(t)$  and  $s_2(t)$  are transmitted over an AWGN channel with the noise power spectral density of  $N_0/2$ , the error probability can be expressed as

$$P_e = Q\left(\sqrt{\frac{d^2}{2N_0}}\right),$$

where  $d^2 = \int_{-\infty}^{\infty} (s_1(t) - s_2(t))^2 dt$ . Now, we further consider the case that  $s_1(t) = u(t)$  and  $s_2(t) = u(t-1)$ , where  $u(t)$  is shown in Fig. 1. Answer the following questions.

- (a). (10%) Plot the diagram of an optimal matched filter receiver for the system. Explicitly label the required parameters.
- (b). (5%) Decide the error probability for the system.
- (c). (10%) Consider  $s_1(t) = u(t)$  and  $s_2(t) = \begin{cases} u(t-1) & \text{with probability 0.5} \\ u(t) & \text{with probability 0.5} \end{cases}$ . Determine the optimum detection rule.

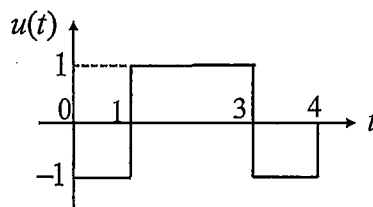


Fig. 1

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4. (20%) A carrier-modulated signal of  $x(t) = a(t) \cos(\omega_c t + \theta(t))$  can be transformed into an equivalent baseband signal, as shown in Fig. 2.

If  $x(t) = \text{sinc}(10^5 t) \cos(2\pi 10^6 t) + \text{sinc}(10^5 t) \sin(2\pi 10^6 t)$ , answer the following questions.

(a). (10%) Decide the values of  $\omega_c$ ,  $a(t)$ , and  $\theta(t)$ .

(b). (10%) Determine  $x_b(t)$  and  $x_A(t)$

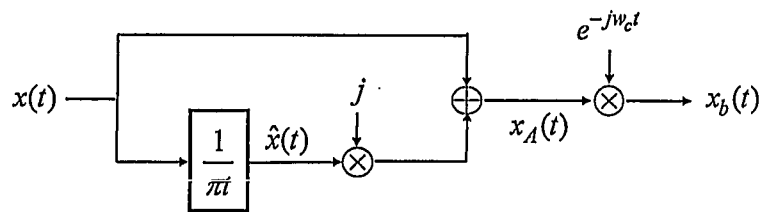


Fig. 2. Transform a passband signal to a baseband signal.

5. (15%) For a binary communication system shown in Fig. 3, the receiver obtains the two corrupted signals,  $y_1$  and  $y_2$ , where two noises,  $n_1$  and  $n_2$ , are not necessarily Gaussian distributions. The maximum a posteriori probability (MAP) receiver can be used to optimally detect the transmitted signal  $s$  from the observed signals,  $y_1$  and  $y_2$ , i.e.,  $\hat{s} = \max_s p(s | y_1, y_2)$ . If  $n_1$  and  $n_2$  are independent, can the optimum decision be based only on  $y_1$  (i.e.,  $\hat{s} = \max_s p(s | y_1, y_2) = \max_s p(s | y_1)$ )? Please justify your answer.

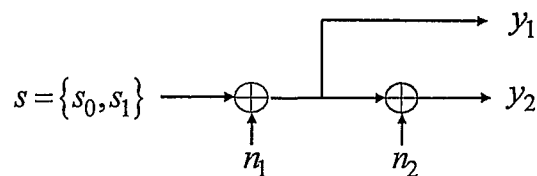


Fig. 3.

6. (20%) Denote  $x(t)$  and  $y(t)$  as the two bandpass real signals, and  $x_L(t)$  and  $y_L(t)$  are the corresponding lowpass equivalents with respect to the carrier frequency  $f_0$ . Thus, they have the

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following relations in frequency domain:

$$X(f) = \frac{1}{2} X_L(f - f_0) + \frac{1}{2} X_L^*(-f - f_0)$$

and

$$Y(f) = \frac{1}{2} Y_L(f - f_0) + \frac{1}{2} Y_L^*(-f - f_0),$$

where  $X(f)$ ,  $Y(f)$ ,  $X_L(f)$ , and  $Y_L(f)$ , are the frequency responses of  $x(t)$ ,  $y(t)$ ,  $x_L(t)$ , and  $y_L(t)$ , respectively.

- (a). (15%) Assume  $X_L(f - f_0)$  and  $Y_L(-f - f_0)$  do not overlap. Show that  $\int_{-\infty}^{\infty} x(t)y(t)dt = \frac{1}{2} \text{Re} \left\{ \int_{-\infty}^{\infty} x_L(t)y_L^*(t)dt \right\}$ , where  $\text{Re}\{X\}$  represents the real part of the complex number  $X$ . Hint: Use Parseval's relation  $\int_{-\infty}^{\infty} x(t)y(t)dt = \int_{-\infty}^{\infty} X(f)Y^*(f)df$ .
- (b). (5%) Prove that the energy in a bandpass signal is just one-half the energy in its lowpass equivalent.